

Chapter 14 Introduction to Microwave Systems

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radiation pattern, directivity, efficiency, gain, temperature
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14.1 System aspects of antennas

- antenna characteristics

radiation pattern $F(\theta, \phi) \equiv r^2 |\vec{E}(\theta, \phi) \times \vec{H}^*(\theta, \phi)| = r^2 S(\theta, \phi)$

directivity $D \equiv \frac{4\pi F_{\max}}{P_{rad}} = \frac{4\pi F_{\max}}{\iint F(\theta, \phi) \sin \theta d\theta d\phi} = \frac{4\pi A_e}{\lambda^2} \quad A_e \equiv \frac{P_l}{S}$

efficiency $\eta \equiv \frac{P_{rad}}{P_{in}}$

gain $G = \eta D$

equivalent noise temperature

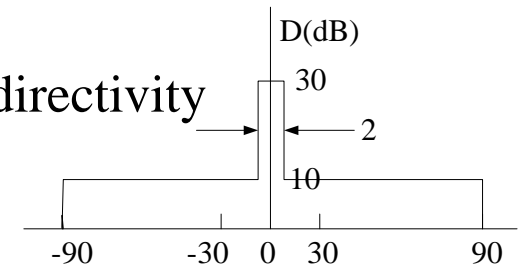
$$T_A = \eta T_b + (1 - \eta) T_p$$

$$T_b = \frac{\iint T_B(\theta, \phi) D(\theta, \phi) \sin \theta d\theta d\phi}{\iint T_B(\theta, \phi) \sin \theta d\theta d\phi}$$

Discussion

1. Ex.14.3 Calculate T_A with $\eta=1$ and an antenna directivity

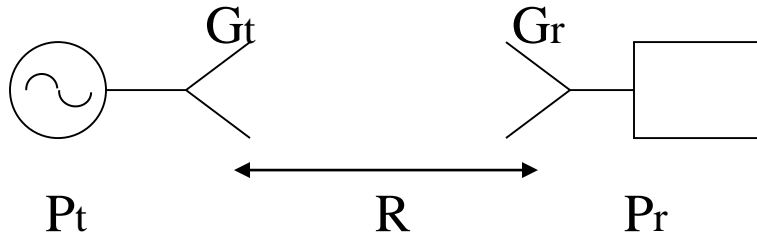
$$\text{background noise temperature } T_B = \begin{cases} 10^\circ K, & |\theta| \leq 30^\circ \\ 100^\circ K, & 30^\circ < |\theta| \leq 90^\circ \end{cases}$$



$$\rightarrow T_A = T_b = \frac{\int_0^{1^\circ} 10 \times 1000 \sin \theta d\theta + \int_{1^\circ}^{30^\circ} 10 \times 10 \sin \theta d\theta + \int_{30^\circ}^{90^\circ} 100 \times 10 \sin \theta d\theta}{\int_0^{1^\circ} 1000 \sin \theta d\theta + \int_{1^\circ}^{90^\circ} 10 \sin \theta d\theta} = 86.4^\circ K \approx \text{sidelobe region}$$

14.2 Microwave communication systems

- Friis power transmission formula



$$P_r = \frac{P_t G_t A_e}{4\pi R^2} = \frac{P_t G_t G_r \lambda^2}{(4\pi R)^2}$$

Discussion

1. Ex.14.4 a DBS satellite @12.45GHz, $P_t=120W$, $G_t=34dB \rightarrow 2m$ dia. $\theta=1.7^\circ$, slant range 39000km, ground receiving terminal $G_r=33.5dB \rightarrow 18in$ dia. $\theta=6^\circ$, $T_A=50K$, $NF=0.7dB$, $BW=20MHz$

(1) transmitter $EIRP = P_t G_t = 120 \times 2512 = 54.8dBm$

(2) receiver $T_e = T_A + T_{LNB} = T_A + (F - 1)T_o = 50 + (1.179 - 1) \times 290 = 100.8K$

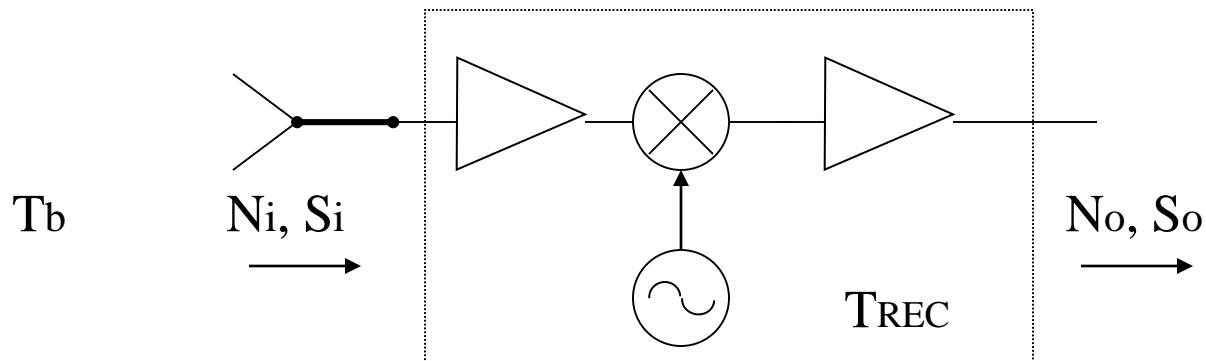
receiver $\frac{G}{T} = \frac{2239}{100.8} = 13.5dB/K$

(3) $P_r = \frac{P_t G_t G_r \lambda^2}{(4\pi R)^2} = -117.9dBW = -87.9dBm$

(4) $CNR = \frac{P_r G_{LNB}}{kT_e B G_{LNB}} = 17.7dB > \text{system link margin}$

- noise analysis of a microwave receiver

GA, η , T_p L_T , T_p G_{RF} , T_{RF} L_M , T_M G_{IF} , T_{IF}



equivalent noise temperature

(1) antenna

$$T_A = \eta T_b + (1 - \eta) T_p, N_i = k T_A B$$

(2) cable

$$T_{TL} = (L_T - 1) T_p$$

(3) receiver

$$T_{REC} = T_{RF} + \frac{T_M}{G_{RF}} + \frac{T_{IF} L_M}{G_{RF}} \rightarrow T_{TL+REC} = T_{TL} + L_T T_{REC}$$

\Rightarrow

$$\text{o/p signal power } S_o = \frac{S_i G_{RF} G_{IF}}{L_T L_M} = S_i G_{sys}$$

o/p noise power

$$\begin{aligned} N_o &= (N_i + kBT_{TL+REC}) G_{sys} \\ &= kB(T_A + T_{TL+REC}) G_{sys} \\ &= kB[\eta T_b + (1 - \eta) T_p + (L_T - 1) T_p + L_T T_{REC}] G_{sys} \\ &= kBT_{sys} G_{sys} \end{aligned}$$

$$SNR_o = \frac{S_o}{N_o} = \frac{S_i}{kBT_{sys}}$$

Discussion

1. Ex14.5 a 4 GHz receiver with BW=1MHz, $S_i = -80\text{dBm}$

$$\eta = 0.9, T_b = 200^\circ\text{k}, T_p = 300^\circ\text{k} \rightarrow T_A = 210^\circ\text{k}$$

$$\rightarrow N_i = kT_A B = -115\text{dBm} \Rightarrow \text{SNR}_i = 35\text{dB}$$

$$L_T = 1.5\text{dB} = 1.41 \rightarrow T_{TL} = (L_T - 1)T_o = 123^\circ\text{k},$$

$$L_M = 6\text{dB} = 4, F_M = 7\text{dB} \rightarrow T_M = (F_M - 1)T_o = 1163^\circ\text{k},$$

$$F_{RF} = 3\text{dB} \rightarrow T_{RF} = (F_{RF} - 1)T_o = 289^\circ\text{k}$$

$$F_{IF} = 1.1\text{dB} \rightarrow T_{IF} = (F_{IF} - 1)T_o = 84^\circ\text{k}$$

$$G_{RF} = 20\text{dB} = 100, G_{IF} = 30\text{dB} = 1000$$

$$\rightarrow T_{REC} = T_{RF} + \frac{T_M}{G_{RF}} + \frac{T_{IF} L_M}{G_{RF}} = 304^\circ\text{k}$$

$$\rightarrow T_{sys} = T_A + T_{TL} + L_T T_{REC} = 762^\circ\text{k} \rightarrow kBT_{sys} = -110\text{dBm}$$

$$\Rightarrow \text{SNR}_o = \frac{S_i}{kBT_{sys}} = -80 + 110 = 30\text{dB}$$

- noise analysis of a digital communication receiver

$$P_b : BER\left(\frac{E_b}{n_o} \equiv \frac{\text{bit energy}}{\text{noise power density}}\right)$$

$$\frac{E_b}{n_o} = \frac{ST_b}{n_o} = \frac{S}{n_o R_b} = \frac{S}{\frac{N}{B} R_b} = \frac{SB}{NR_b}, S : \text{signal power}, T_b = \frac{1}{R_b} : \text{bit period}, R_b : \text{bit rate}$$

Ex14.6 a LEO communication satellite $f=16\text{GHz}$, $P_t=80\text{W}$, $G_t=20\text{dB}$, $R=940\text{km}$, a QPSK receiving handset $G_r=1\text{dB}$, $T_{\text{sys}}=750\text{K}$, $L_{\text{atm}}=2\text{dB}$ $BER=0.01$, find $R_{b,\text{min}}$ for 10dB link margin

$$P_r = \frac{P_t G_t G_r \lambda^2}{(4\pi R)^2} = 49 + 20 + 1 - 176 - 2 = -108\text{dBm}$$

$$S_{\text{min}} = -108 - 10 = -118\text{dBm} = 1.58 \times 10^{-15}\text{W}$$

$$\text{QPSK } BER = 0.01 \rightarrow \frac{E_b}{n_o} = 5\text{dB} = 3.16 = \frac{S_{\text{min}} B}{NR_{b,\text{max}}}$$

$$\rightarrow R_{b,\text{max}} = \frac{n_o}{E_b} \frac{S_{\text{min}} B}{N} = \frac{n_o}{E_b} \frac{S_{\text{min}} B}{kT_{\text{sys}} B} = \frac{n_o}{E_b} \frac{S_{\text{min}}}{kT_{\text{sys}}} = \frac{1}{3.16} \frac{1.58 \times 10^{-15}}{1.38 \times 10^{-23} \times 750} = 48\text{kbps}$$

14.3 Radar systems

- radar (radio detection and and ranging) equation

$$P_r = \frac{P_t G}{4\pi R^2} \sigma \frac{1}{4\pi R^2} A_e = \frac{P_t G^2 \lambda^2 \sigma}{(4\pi)^3 R^4} \rightarrow R_{\max} = \sqrt[4]{\frac{P_t G^2 \lambda^2 \sigma}{(4\pi)^3 P_{\min}}}$$

Discussion

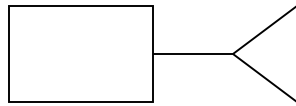
1. Ex14.7 radar @ 10GHz, $P_t=2\text{kW}$, $G=28\text{dB}$, $\sigma=12\text{m}^2$, $P_{\min}=-90\text{dBm}$
→ $R_{\max} = 8114\text{km}$
2. pulse radar → detect target range and direction
3. Doppler radar → detect target radial velocity
4. radar cross section (p.696, Table 14.3)
e.g. bird 0.01m^2 , missile 0.5m^2 , person 1m^2 , fighter $3\text{-}8\text{m}^2$,
airplane 100m^2 , truck 200m^2

5. SIR-C (1994) microwave image of Taipei area

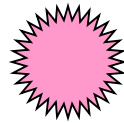


14.4 Radiometry

- radiometer



$$P = kTB B = keTB$$



T

T: physical temperature

T_B : brightness temperature

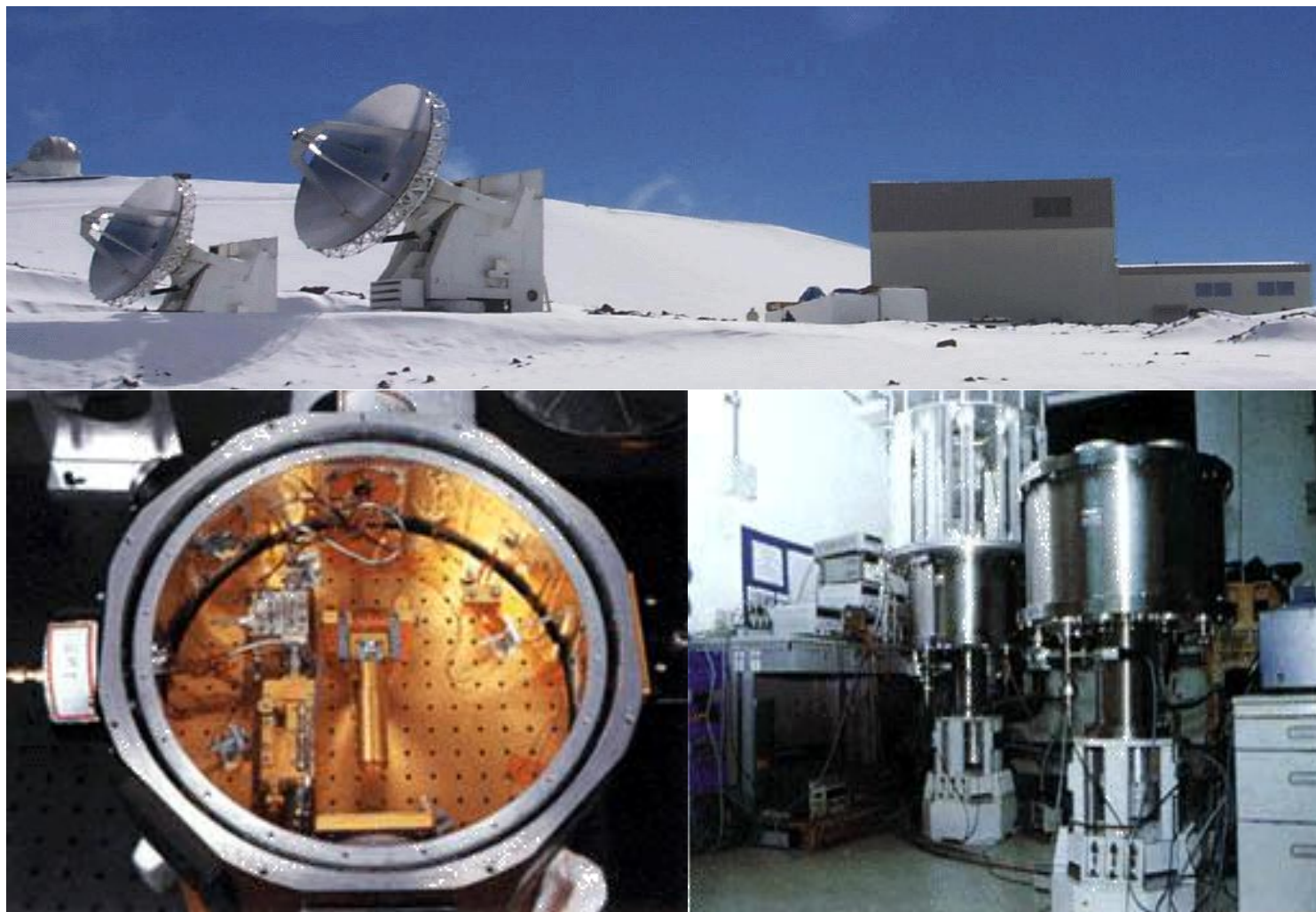
e: emissivity

$$0 \leq e \leq 1$$

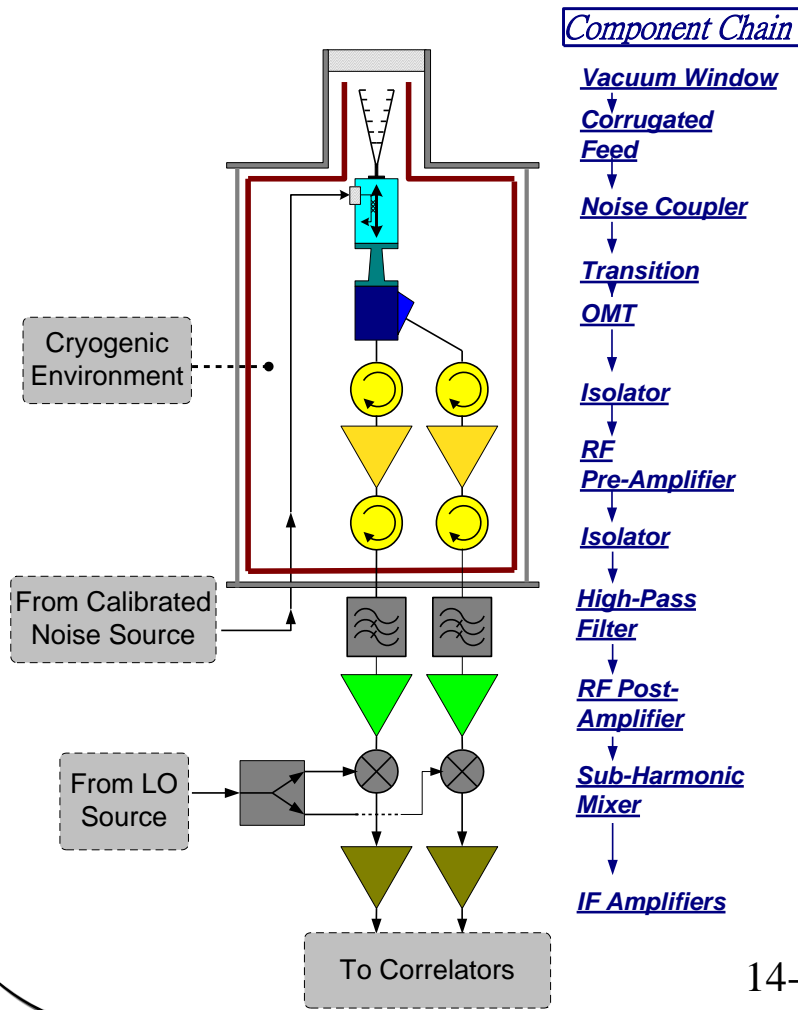
Discussion

1. radiometer measures the object noise power $\rightarrow T_B \rightarrow$ object physical parameters (p.699)
e.g. soil moisture, ocean surface wind speed, target image, mapping of galactic objects,.....
2. usually $T_B < T_R$ receiver temperature, $V_o = Gk(T_B + T_R)B$,
 T_R can be derived from the Y-factor calibration
3. measurement errors due to
noise fluctuation \rightarrow increase integration time
receiver gain fluctuation \rightarrow Dicke radiometer

4. SMA (sub-millimeter array) (1983, 1996~) 200GHz-900GHz



5. AMiBA (Array for Microwave Background Anisotropy) 宇宙背景輻射陣列望遠鏡(2001~) 85GHz-105GHz



14.5 Microwave propagation

Discussion

1. “line of sight propagation” , $1/R^2$, for microwaves
→ satellite communication
2. atmospheric attenuation effects (p.703, Fig.14.29)
negligible for $f < 10\text{GHz}$
water vapor resonance @ 22.2GHz, 183.3GHz
O₂ resonance @ 60GHz, 120GHz
“microwave window” @ 35GHz, 94GHz, 135GHz
→ earth remote sensing
3. ground reflections
→ fading or scintillation problems for wireless communication
in urban area
→ clutter problem for radar
4. plasma frequency $\cong 8\text{MHz}$, total reflection from ionsphere
for $f < f_p$ → short wave radio

14.6 Other applications and topics

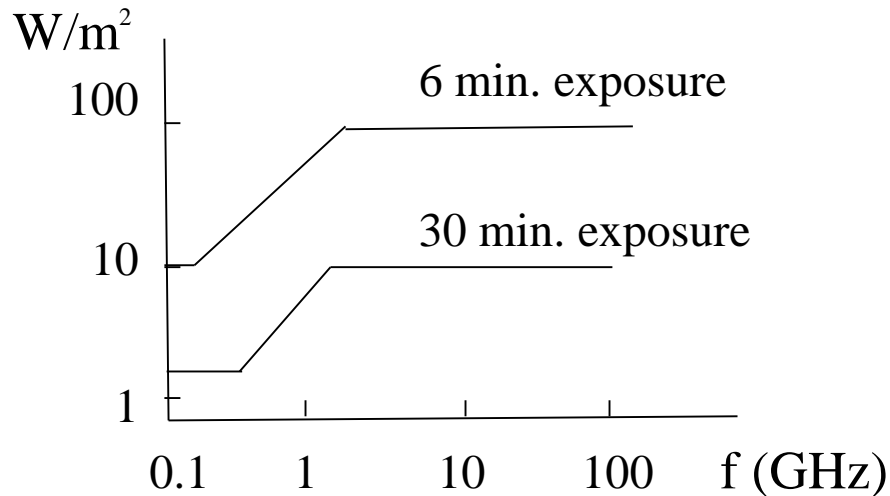
Discussion

1. microwave oven @ 2.45GHz

→ conduction loss for large $\tan\delta$ of material

2. energy transfer → nonpilot airplane, solar power satellite station, mobile phone, ...

3. microwave exposure → thermal effects, IEEE Standard C95.1-2005
power density: $10\text{W/m}^2 \sim 1\text{mW/cm}^2$



4. Ex 14.8 a 18GHz microwave antenna 36dB gain and 10W radiation
→ power density at 20 m away

$$S = \frac{P_{in}G}{4\pi R^2} = \frac{10 \times 4000}{4\pi \times 20^2} = \begin{cases} 8W/m^2 @ main lobe \\ 0.8W/m^2 @ -10dB sidelobe \end{cases} < 10W/m^2$$

ADS examples: Ch14_prj